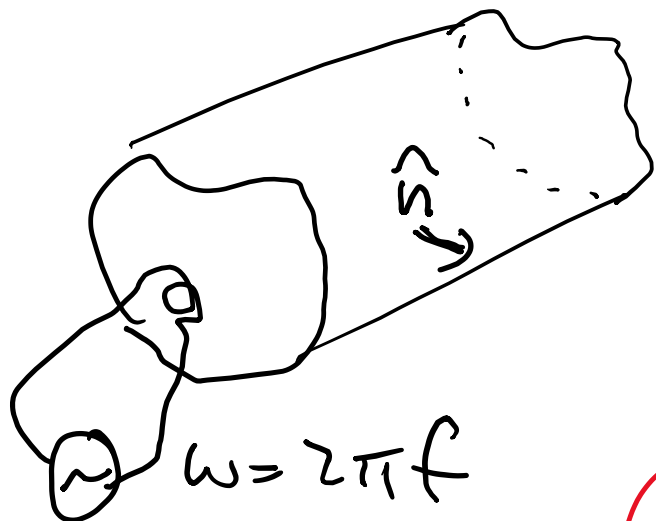


#7: RF CAVITIES

1/24/19



WAVEGUIDE:

- open at both ends
- constant (arbitrary) cross-section
- perfectly conducting

TRANSMITTING (?) fields oscillating at $\omega = 2\pi f$

Boundary conditions at the walls:

$$\hat{n} \times \vec{E} = 0 \quad \text{no parallel } E\text{-fields}$$

$$\hat{n} \cdot \vec{B} = 0 \quad \text{no perpendicular } B\text{-fields}$$

\hat{n} - unit vector perpendicular to wall

Real parts of complex \vec{B} & \vec{E} vectors are physical

Consider a mode labeled by WAVE NUMBER k

(A)

$$\vec{E} = \vec{E}(r, \theta) \cdot e^{i(kz - \omega t)}$$

$$\vec{B} = \vec{B}(r, \theta) \cdot e^{i(kz - \omega t)}$$

where phase velocity in z -direction is

$$v_p = \omega / k$$

If k is imaginary then wave is damped

DOES NOT PROPAGATE

Q: how does k vary with ω ?

Q: WHAT happens when walls are added at waveguide ends, making a (Cavity)

TRANSVERSE MODES

3 categories of mode solve

(A)

1 TRANSVERSE MAGNETIC

(TM)

$B_z = 0$ everywhere, $E_z = 0$ at walls
 $\neq 0$ in general.

2 TRANSVERSE ELECTRIC

(TE)

$E_z = 0$ everywhere, with $\frac{\partial B_z}{\partial u} = 0$ at walls

TM modes are most useful: accelerate/decelerate
(but crab cavities (TE) are state-of-art ^{beam})

3 TRANSVERSE ELECTROMAGNETIC

(TEM)

NO longitudinal fields: FREE-SPACE wave

$$k = \sqrt{\mu_0 \epsilon_0} \frac{\omega}{c}$$

In a vacuum $\mu_R = \epsilon_R = 1$ = TEM modes
propagate @ "c" "UNAWARE" of walls because
their wavelengths $\lambda = \frac{2\pi}{k}$ are much smaller
than waveguide width.

SOLVE FOR TM (or TE) MODES

of a particular geometry identifies a family
of cut-off frequencies ω_n with $n = 0, 1, \dots, \infty$

$$k_n = \pm \frac{1}{c} \sqrt{\omega^2 - \omega_n^2}$$

- Clearly mode n does NOT propagate if $\omega < \omega_n$
- $\pm k$ modes propagate forwards or backwards
(or both)

- Many modes n can co-exist if ω is large enough

- It is advantageous to:

1) Drive at $\omega_0 \ll \omega \ll \omega_1$ ONE MODE

2) Use a geometry with $\omega_0 \ll \omega_1$ DYNAMIC RANGE

- The beam ITSELF can drive many unwanted HIGHER ORDER MODES (HOM's) up to frequencies

$$\omega_{\text{MAX}} \approx 2\pi \frac{c}{\sigma_s}$$

where RMS bunch length may be (is) short.

- HOM's sometimes need to be explicitly damped...

PILL BOXES - CYLINDRICAL RESONANT CAVITIES

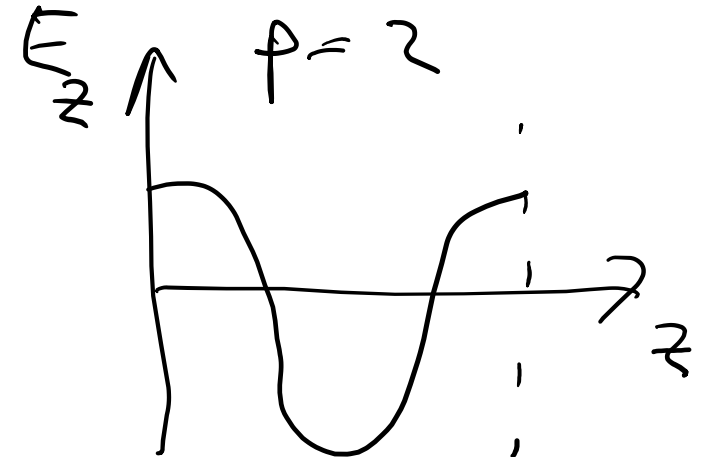
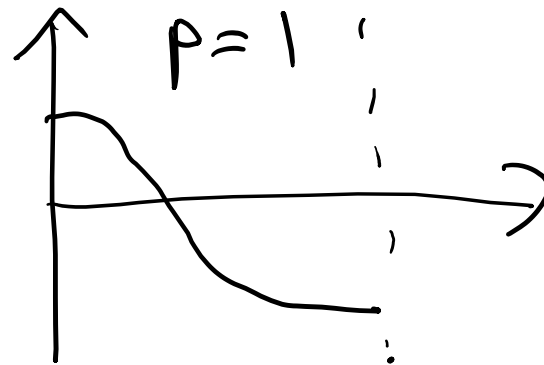
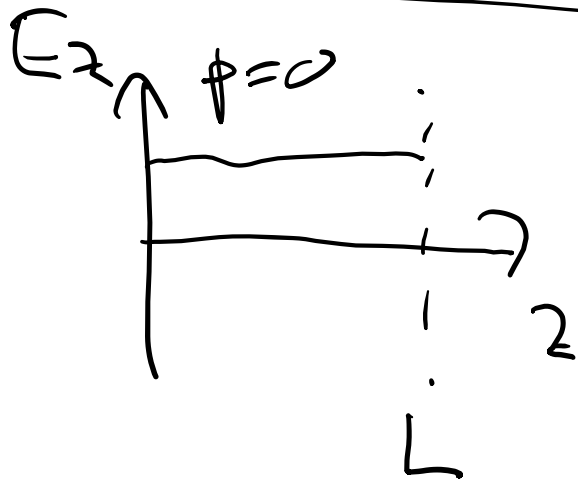
ADD FLAT ENDS at $z=0$ & L

Take a pair of $\pm k$ modes with frequency ω

$$k = \pm k_p = \pm p \frac{\pi}{L} \quad p=0, 1, \dots, \infty$$

and add together to form a TM mode

$$E_z = \psi(r, \theta) \cdot \cos\left(p\pi \frac{z}{L}\right) \cdot e^{-i\omega t} \quad \text{RES}$$



NEXT: SOLVE $\psi(r, \theta)$ for a circle

- adds 2 more indices: m, n

where the natural resonant frequencies for a pill-box are

$$\textcircled{A} \quad \omega_{mnp} = c \sqrt{\left(\frac{U_{mn}}{R}\right)^2 + \left(\frac{p\pi}{L}\right)^2} \quad [s^{-1}]$$

and $U_{mn} \geq 1$ is with root of with Bessel function

- Lowest resonant frequency has $p = 0$

- Simplify further with $m = 0$
(no azimuthal structure)

TM_{0n0} MODE has only 2 non-zero field components

$$\begin{aligned}
 E_z &= E_0 J_0\left(u_{0n} \frac{r}{R}\right) e^{-i\omega_{0n0} t} \\
 B_\theta &= -B_0 \frac{\omega_{0n0} R}{u_{0n} c} J_1\left(u_{0n} \frac{r}{R}\right) e^{-i\omega_{0n0} t}
 \end{aligned}$$

As illustrate (in book) for TM₀₂₀

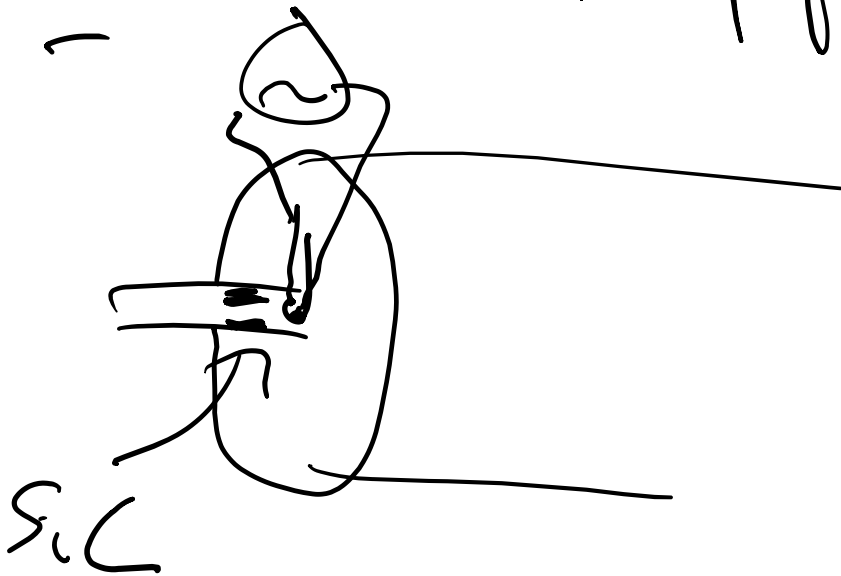
- n counts the E_z zero crossing between r = 0 & R
- Note that B_z is non-zero at r = R

$$\omega_{0n0} = \frac{u_{0n} c}{R}$$

$$\begin{pmatrix} u_{01} \\ u_{02} \\ u_{03} \end{pmatrix} = \begin{pmatrix} 2.41 \\ 5.52 \\ 8.65 \end{pmatrix}$$

ADD HOLES to let beam through ...

- This let's field test a little way into the beam pipe, but not far if ω_{hole} is much smaller than the pipe cut-off frequency
- if pipe radius $\ll R$



ADJUST Q to get the right frequency: (see (A))

How LONG should L be?

- A particle with (constant) speed βc passes through center at $t=0$, acquires a voltage

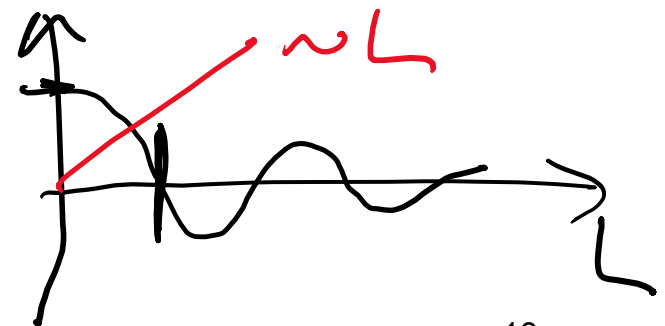
$$V_A = \int_{-L/2}^{L/2} E_z \cdot dz = \beta c \cdot E_0 \int_{-L/2\beta c}^{+L/2\beta c} e^{i\omega t} \cdot dt$$

With TM_{010} mode, then

$$V_A = E_0 L \cdot T_1$$

TRANSIT TIME FACTOR

$$T_1(L) = \frac{\sin(\omega L / 2\beta c)}{(\omega L / 2\beta c)}$$



and V_A has a maximum when

$$T = 2/\pi$$

$$L_{\text{opt}} = \pi \frac{\beta c}{\omega}$$

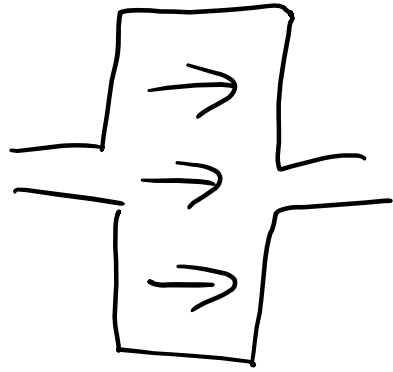
EG in TM_{010} $L_{\text{opt}} = \frac{\pi}{2.41} \beta R$

VERY INEFFICIENT when $\beta \ll 1$!! Protons or ions
NON-RELATIVISTIC particles (say $\beta \leq 0.5$) need
different cavity geometries.

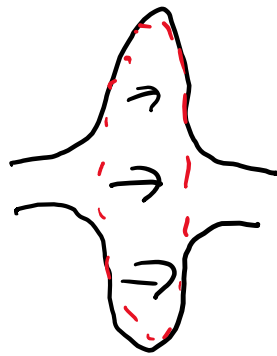
RESONANT ZOO:

Spike, split-ring, inter-digital, $1/4$ wave
 $1/2$ -wave, ... - - RFQ, ...

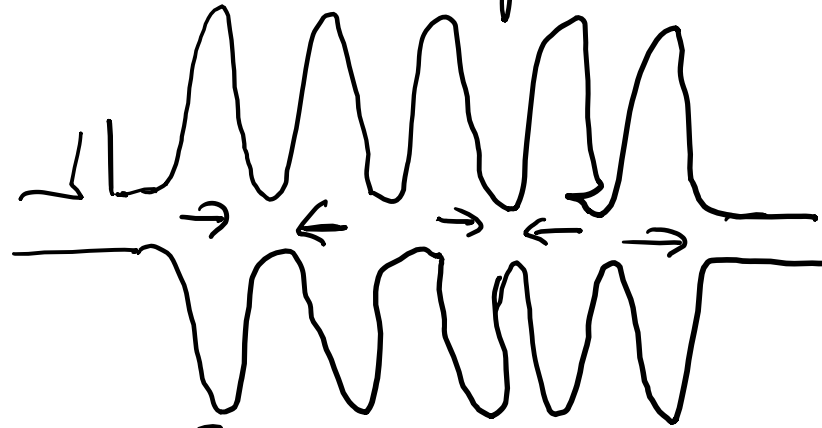
RELATIVISTIC CAVITIES: Geometries morph away from pill-boxes:



PILL BOX



ELLIPTICAL
1-CELL



ELLIPTICAL N-CELL
with π -phase shift

The TM_{mnp} "language" survives if rotational symmetry is preserved

Q1: WHAT is the optimum # of cells in a cavity?
(see later lecture)

Q2: How BIG can gradient E_0 be? Copper cavities vs. superconducting?

KILPATRICK CRITERION (1950's)

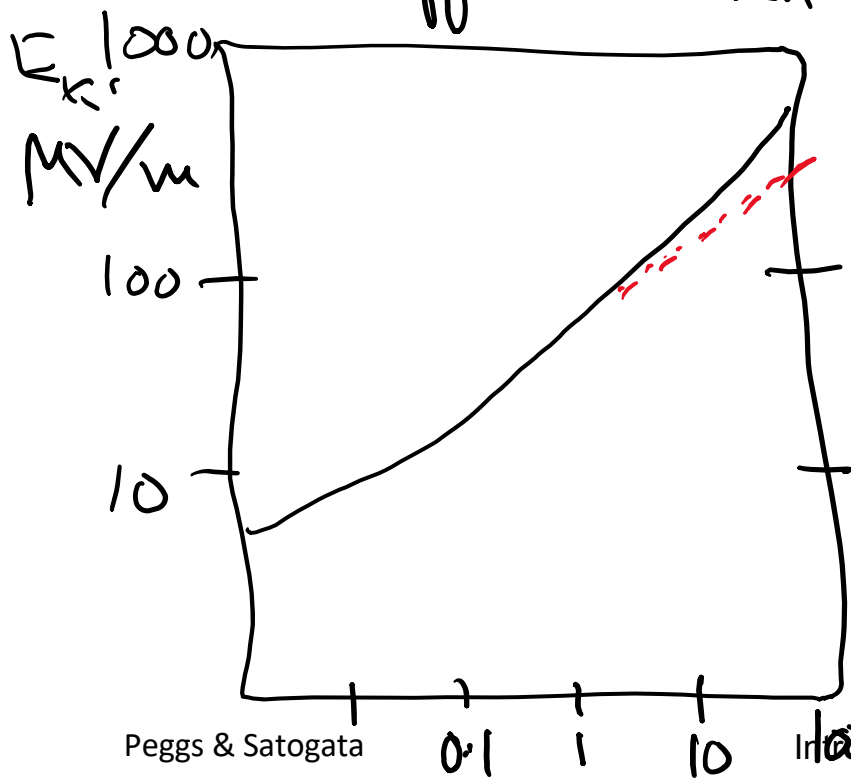
Empirically:

$$f = 1.64 E_k^2 e^{-(8.5/E_k)}$$

↑
↑

MHz
 MV/m

defines the frequency necessary to reach a gradient E_k on a copper wall without breakdown



- Conservative by ~ a factor of 2
- Suggests (falsely?) that copper cavities at $f > 10$ GHz can outstrip superconducting cavities

$(E_{sc} \gg 10 \text{ MV/m})$

SUPERCONDUCTING CAVITIES

- Many geometries are possible despite more difficult manufacture
- Lower operating costs: "no" heat dissipation in conducting walls
- Higher capital cost: complexity, material treatment, cryogenics...
- Very high Q values mean very small bandwidths

$$f_{BW} = \frac{f}{Q} \sim \frac{10^9}{10^{10}} \text{ [Hz]} \Rightarrow 0.1 \text{ [Hz]}$$

and sensitivity to mechanical disturbances:

- 1) microphonics (any noise, e.g. cryogenics)
 - 2) Lorentz force detuning (in pulsed operation)
- \Rightarrow CW operations biases (more) towards SC technology.